



**COLLEGE OF ENGINEERING  
PUTRAJAYA CAMPUS  
FINAL EXAMINATION  
SEMESTER 2 2011 / 2012**

PROGRAMME : Bachelor of Electrical & Electronics Engineering (Honours)  
Bachelor of Electrical Power Engineering (Honours)

SUBJECT CODE : EEEB273

SUBJECT : ELECTRONIC ANALYSIS AND DESIGN II

DATE : January 2012

TIME : 3 hours (9.00 am – 12.00 pm)

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**INSTRUCTIONS TO CANDIDATES:**

1. This question paper contains **SIX (6)** questions in **TWELVE (12)** pages.
2. Answer **ALL** questions.
3. Write **all** answers in the answer booklet provided.
4. Write answer to each question on **a new page**.
5. For all calculations, assume that  $V_T = 26 \text{ mV}$ .

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***THIS QUESTION PAPER CONSISTS OF TWELVE (12) PRINTED PAGES INCLUDING  
THIS COVER PAGE AND APPENDIX.***

**Question 1 [16 marks]**

(a) A basic **two-transistor BJT current source** consists of two **npn-type BJT transistors** and a resistor,  $R_1$ , to establish its **reference current,  $I_{REF}$** . The reference current is established when the circuit is connected to positive and negative power source,  $V^+$  and  $V^-$ .

(i) Let the second transistor,  $Q_2$ , be the output transistor. **Draw and label** the circuit and its components **clearly**.

[3 marks]

(ii) Assume the transistors are matched and the transistors parameters are  $\beta = 88$ ,  $I_{S1} = 4.5 \times 10^{-15}$  A, and  $I_{S2} = 3 \times 10^{-15}$  A. If  $I_{REF} = 210 \mu\text{A}$ , calculate  $V_{BE1}$ ,  $V_{BE2}$ , and the bias current,  $I_O$ .

[5 marks]

(b) A **cascode N-MOSFET current source** circuit is described by the following: The output current is **1 mA**; The circuit parameters are  $V^+ = 15$  V and  $V^- = -15$  V; All transistors are identical; The transistor parameters are  $V_{DS(\text{sat})} = 0.8$  V,  $V_{TN} = 1.5$  V,  $k'_n = 120 \mu\text{A/V}^2$  and  $\lambda = 0$ .

(i) **Find the aspect ratio** for **all transistors** of the cascode current source.

[4 marks]

(ii) What is the required resistance value,  $R$ , to establish the **reference current**?

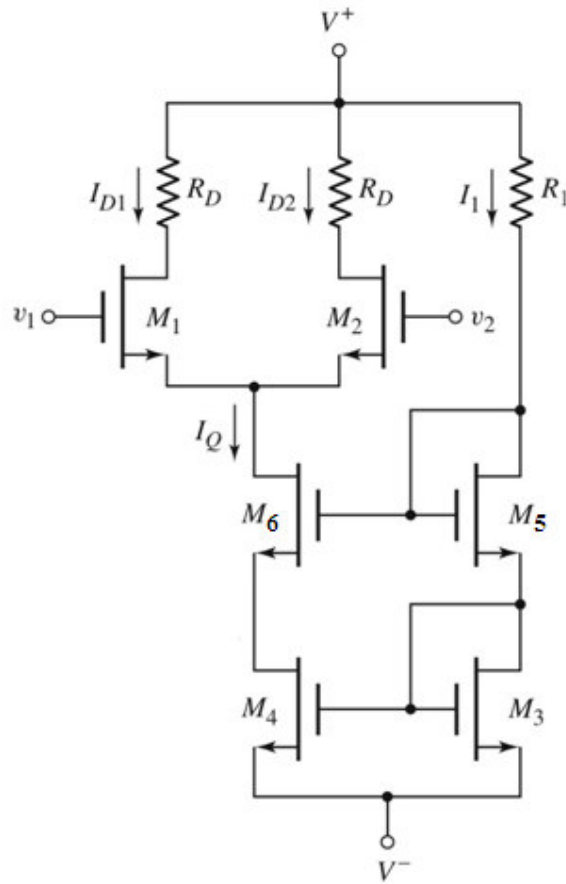
[2 marks]

(iii) **How much** will the output current change if the output voltage changes by **5 V**?

[2 marks]

**Question 2 [16 marks]**

Consider the circuit shown in **Figure 2**, where a MOSFET differential amplifier (consists of  $M_1$  and  $M_2$ ) is biased by a cascode current source (consists of  $M_3, M_4, M_5,$  and  $M_6$ ). The bias voltages are +3 V and -3 V. The transistor parameters are  $K_{n1} = K_{n2} = 150 \mu\text{A}/\text{V}^2$ ,  $K_{n3} = K_{n4} = K_{n5} = K_{n6} = 100 \mu\text{A}/\text{V}^2$ , and  $V_{TN} = 0.3 \text{ V}$  for all transistors.



**Figure 2**

- (a) Calculate  $R_1$  and  $R_D$  in the circuit such that  $V_{DS1} = V_{DS2} = 3 \text{ V}$ ,  $V_{DS3} = V_{DS5}$ , and  $I_{D1} = I_{D2} = 60 \mu\text{A}$  when  $v_1 = v_2 = -0.5 \text{ V}$ . Hints:  $V_G = V_S + V_{GS}$ ,  $V_D = V_S + V_{DS}$ .

[8 marks]

- (b) **One-sided output** ( $V_o$ ) taken at  $V_{D2}$  can be derived using small-signal equivalent circuit to produce

$$V_o = \frac{g_m R_D}{2} V_d - \frac{g_m R_D}{1 + 2g_m R_o} V_{cm}$$

If  $\lambda = 0.01 \text{ V}^{-1}$  for  $M_3$ ,  $M_4$ ,  $M_5$ , and  $M_6$ :

- (i) **Determine** the **output resistance** ( $R_o$ ) of the MOSFET cascode current source.

[2 marks]

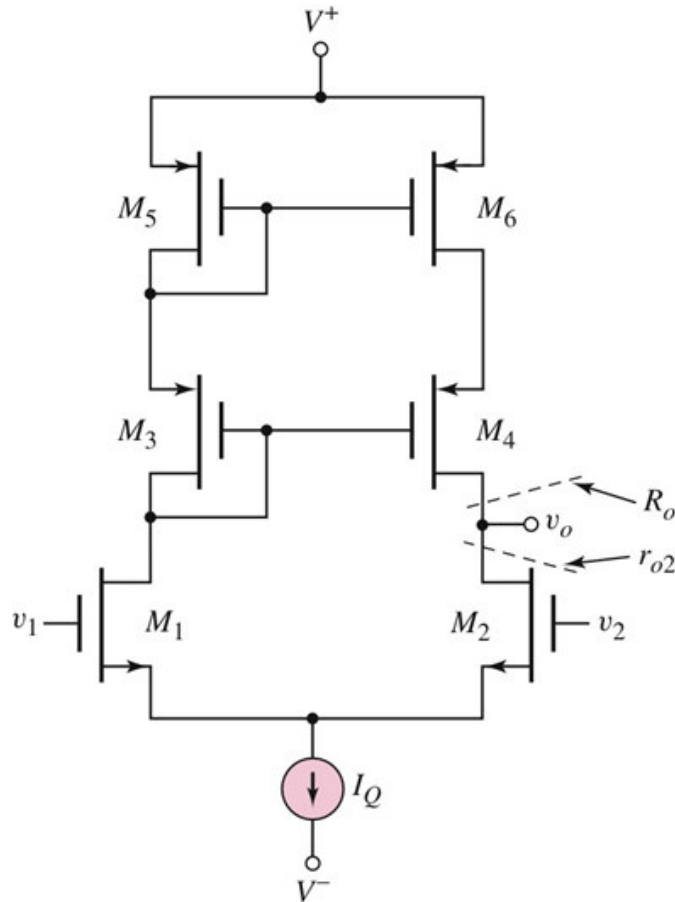
- (ii) **Calculate** the differential-mode voltage gain ( $A_d$ ), common-mode voltage gain ( $A_{cm}$ ), and **CMRR** for the differential amplifier **using result from part (a)**.

[6 marks]

**Question 3 [16 marks]**

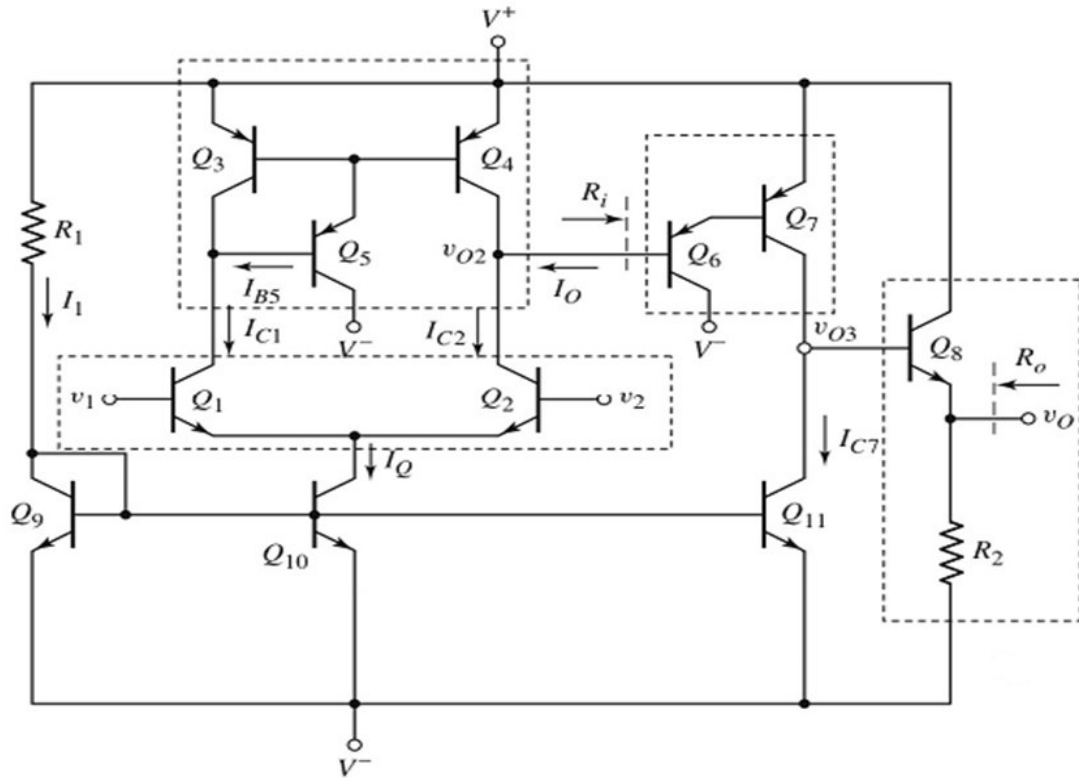
- (a) The differential amplifier in **Figure 3a** has 4 identical PMOS transistors as an **active load**. The circuit is connected to power supply voltages of  $V^+ = +3\text{ V}$  and  $V^- = -3\text{ V}$ . The current source is  $I_Q = 200\text{ }\mu\text{A}$  and its output resistance,  $R_{OCS} = \infty$ . The NMOS transistors parameters are  $V_{TN} = 0.4\text{ V}$ ,  $k'_n = 100\text{ }\mu\text{A/V}^2$ ,  $(W/L)_n = 10$ , and  $\lambda_n = 0.02\text{ V}^{-1}$ . The PMOS transistors parameters are  $V_{TP} = -0.4\text{ V}$ ,  $k'_p = 50\text{ }\mu\text{A/V}^2$ ,  $(W/L)_p = 12$ , and  $\lambda_p = 0.03\text{ V}^{-1}$ . **Determine** the output voltage,  $v_o$ , if the differential input voltage applied is  $v_d = (20 \sin \omega t)\text{ }\mu\text{V}$ .

[8 marks]



**Figure 3a**

(b) The circuit in **Figure 3b** shows a simple multistage BJT op-amp.



**Figure 3b**

(i) **Name** the circuits **indicated** by the boxes.

[2 marks]

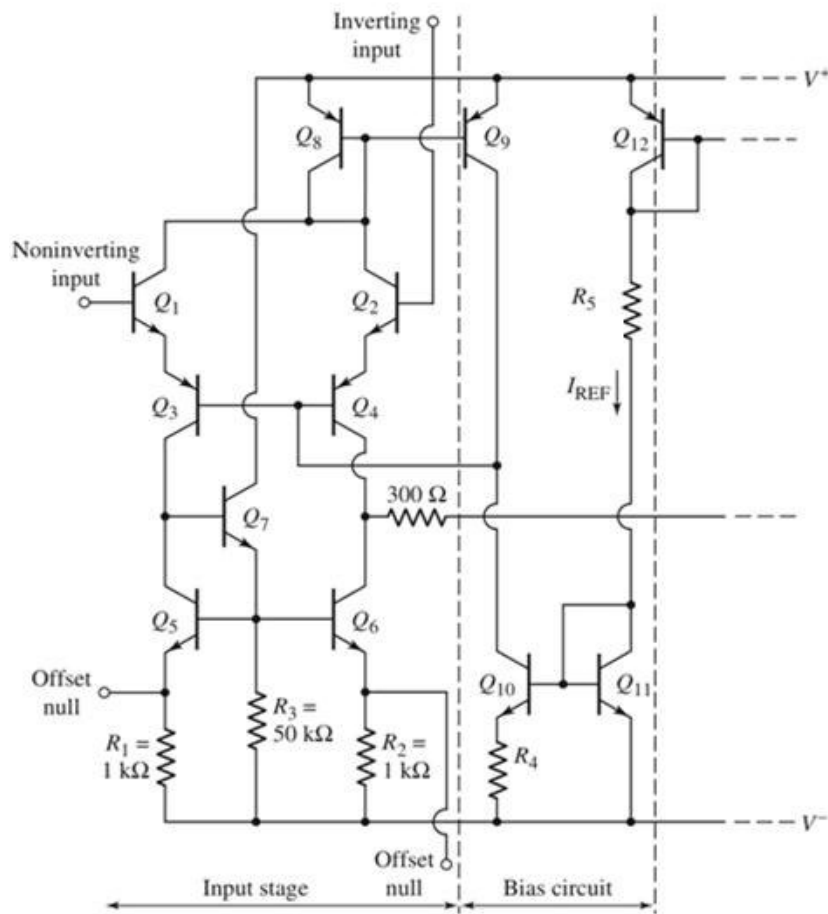
(ii) Assuming that for all transistors:  $r_o = 500 \text{ k}\Omega$ ,  $g_m = 5 \text{ mA/V}$ , and  $r_\pi = 3 \text{ k}\Omega$ , and  $R_2 = 10 \text{ k}\Omega$ . **Calculate** the small signal impedances  $R_i$  and  $R_o$  as indicated in the **Figure 3b**.

[6 marks]

**Question 4 [16 marks]**

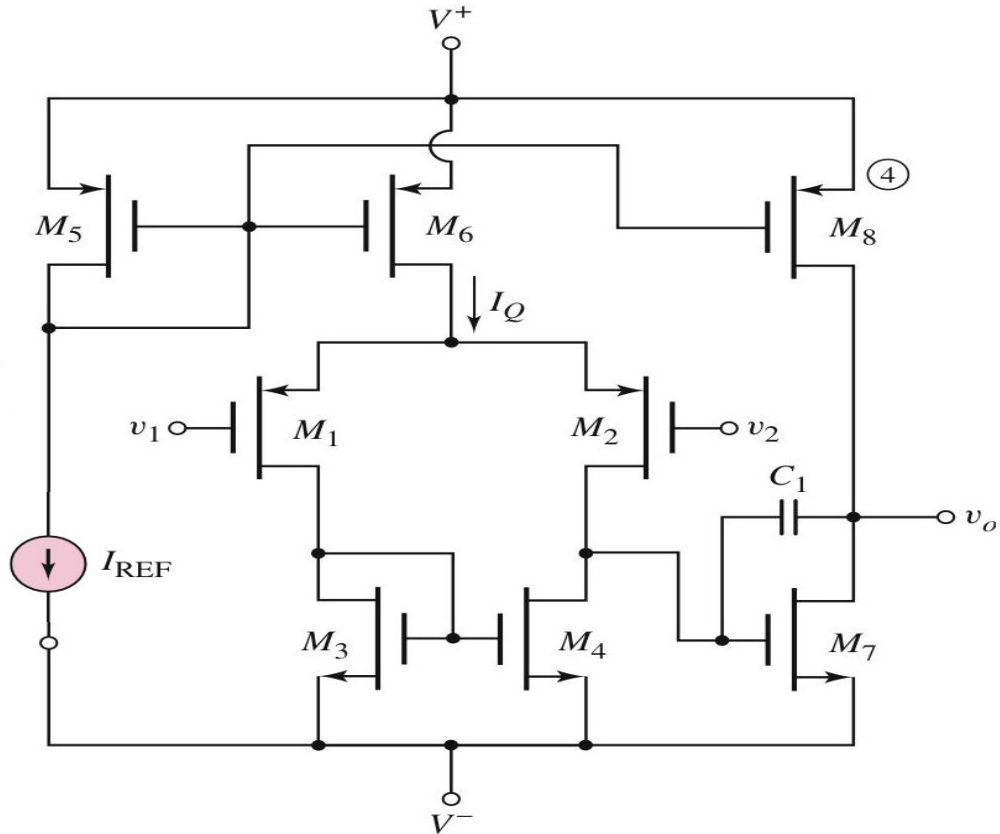
(a) Consider input stage and bias circuit of the 741 operational amplifier in **Figure 4a**, with  $V^+ = +15\text{ V}$ ,  $V^- = -15\text{ V}$ ,  $V_{BE6} = V_{BE7} = 0.6\text{ V}$ ,  $\beta = 200$  and  $I_{C9} = 10\text{ }\mu\text{A}$ . The reverse saturation current  $I_S = 10^{-14}\text{ A}$  for each transistor and the current flow through resistance  $R_5$  is  $1\text{ mA}$ . Ignore the base currents and assume the dc currents in the input stage are exactly balanced.

- (i) Determine the resistance  $R_5$  and  $R_4$ . [4 marks]
- (ii) Find the current  $I_{C1}$  and  $r_{\pi 6}$ . [2 marks]
- (iii) Determine the dc voltage at the collector of  $Q_5$  (i.e.  $V_{C5}$ ). [2 marks]



**Figure 4a**

- (b) Consider the MC14573 op-amp in **Figure 4b**. Assume transistors parameters of  $V_{TN} = 0.5 \text{ V}$ ,  $V_{TP} = -0.5 \text{ V}$ ,  $K_n = 125 \mu\text{A}/\text{V}^2$ ,  $K_p = 100 \mu\text{A}/\text{V}^2$ ,  $V_{SG5} = 1.5 \text{ V}$ ,  $\lambda_n = 0.01 \text{ V}^{-1}$ ,  $\lambda_p = 0.02 \text{ V}^{-1}$  and the circuit parameters of  $V^+ = +10 \text{ V}$  and  $V^- = -10 \text{ V}$ .



**Figure 4b**

- (i) **Find** the dc bias currents  $I_Q$ .

[3 marks]

- (ii) **Determine** the overall voltage gain of the op-amp.

[5 marks]



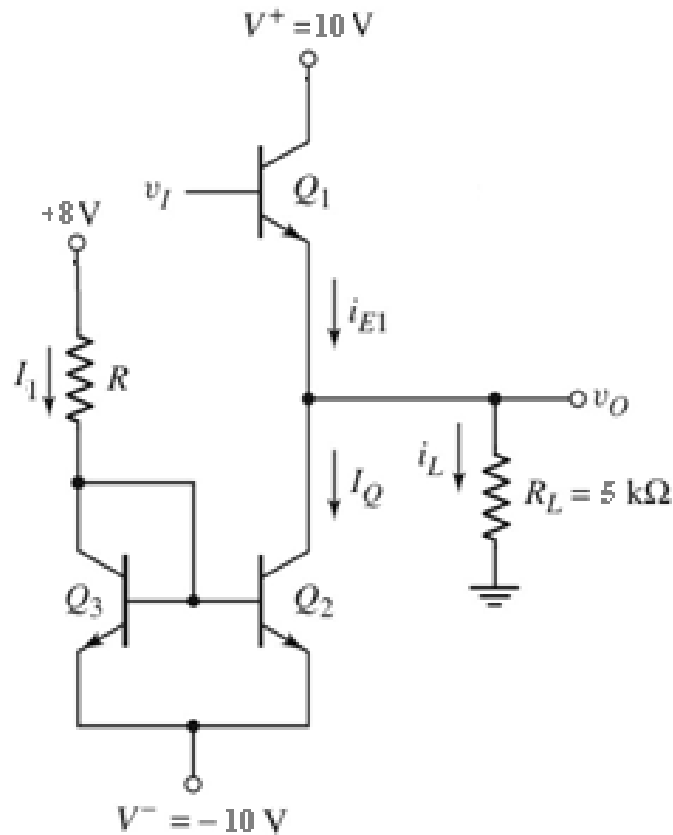
**Question 5 [16 marks]**

(a) Differentiate among class-A, class-B, class-AB, and class-C output stage amplifiers.

[6 marks]

(b) Study **Figure 5** carefully. The transistor parameters are:  $\beta = 180$ ,  $V_{BE(\text{on})} = 0.7 \text{ V}$ , and  $V_{CE(\text{sat})} = 0.4 \text{ V}$ . Design the output stage amplifier for maximum output swing and then calculate its **power conversion efficiency,  $\eta$** .

[10 marks]



**Figure 5**

**Question 6 [20 marks]**

(a) Using **feedback resistor** of **18 kΩ**, draw the following circuits using **inverting op-amp** configuration:

(i) An **inverting amplifier** with a closed-loop gain of **-10**.

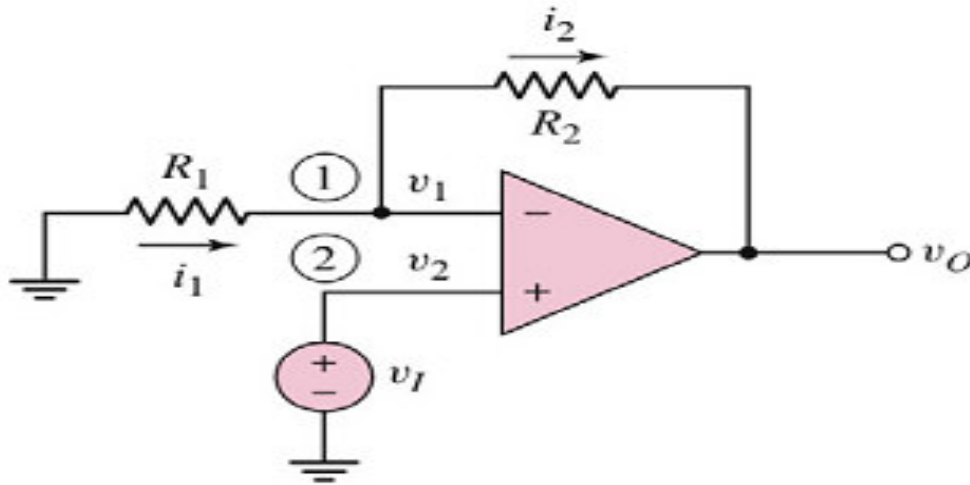
[4 marks]

(ii) A **voltage follower**.

[4 marks]

(b) Consider the **non-inverting op-amp** shown in **Figure 6a**. Assume the op-amp is ideal. Determine the resistor values of  **$R_1$**  and  **$R_2$**  to produce a closed-loop gain of **15**, with the minimum resistance in the circuit is to be **20 kΩ**.

[4 marks]



**Figure 6a**

(c) For a **generalized summing op-amp** as shown in **Figure 6b** the total output voltage is the sum of the individual terms, or

$$v_O = -\frac{R_F}{R_1}v_{I1} - \frac{R_F}{R_2}v_{I2} + \left(1 + \frac{R_F}{R_N}\right)\left(\frac{R_P}{R_A}v_{I3} + \frac{R_P}{R_B}v_{I4}\right)$$

where

$$R_N = R_1 \parallel R_2$$

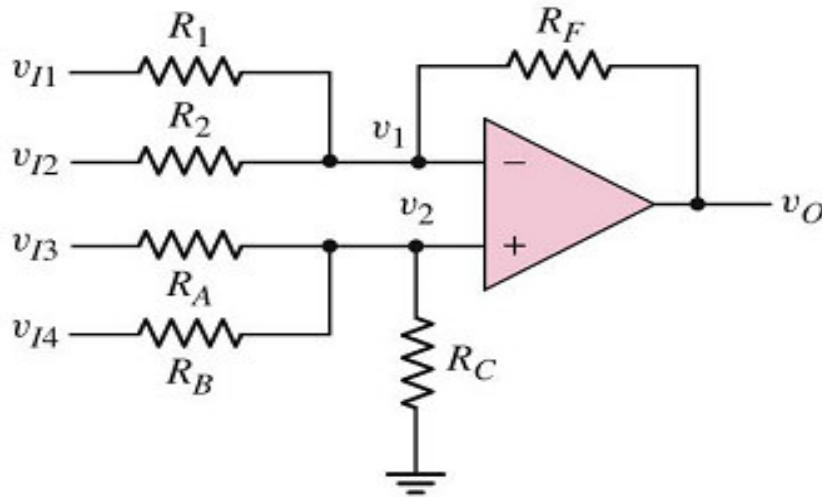
$$R_P = R_A \parallel R_B \parallel R_C$$

**Design** a summing op-amp similar to **Figure 6b** to produce the output

$$v_O = -5v_{I1} - 10v_{I2} + 5v_{I3} + 2v_{I4}$$

The smallest resistor value allowable in the design is **15 kΩ**.

[8 marks]



**Figure 6b**

**-END OF QUESTION PAPER-**

## APPENDIX

BASIC FORMULA FOR TRANSISTORSBJT

$$i_C = I_S e^{v_{BE}/V_T}$$

$$i_C = \alpha i_E = \beta i_B$$

$$i_E = i_B + i_C$$

$$\alpha = \frac{\beta}{\beta + 1}$$

$$r_\pi = \frac{\beta V_T}{I_{CQ}}$$

$$g_m = \frac{I_{CQ}}{V_T}$$

$$\beta = g_m r_\pi$$

$$r_o = \frac{V_A}{I_{CQ}}$$

MOSFET

$$v_{DS}(\text{sat}) = v_{GS} - V_{TN}$$

$$i_D = K_n [v_{GS} - V_{TN}]^2$$

$$K_n = \frac{k'_n}{2} \cdot \frac{W}{L}$$

$$v_{SD}(\text{sat}) = v_{SG} + V_{TP}$$

$$i_D = K_p [v_{SG} + V_{TP}]^2$$

$$K_p = \frac{k'_p}{2} \cdot \frac{W}{L}$$

$$g_m = 2\sqrt{K_n I_{DQ}}$$

$$r_o \cong \frac{1}{\lambda I_{DQ}}$$